

ECE 488 – Automatic Control

Bode Plot Design

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Compulsory Course in Electronic and Communication
Engineering
Credits (3/0/3)

Course Webpage: <http://ECE488.cankaya.edu.tr>

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Reminder

Reminder

Previous Weeks

- LTI system modeling
- Nonlinear modeling and linearization
- Stability
- Steady-state and transient response
- Feedback Control
 - Root locus
 - Nyquist plot
- Bode plot

This week

- Bode plot design
- Lead-lag compensation

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Closed-loop Requirements: Motivation

Open-loop Transfer Function

$$G_o(s) = C(s) G(s)$$

Complementary Sensitivity

$$T(s) = \frac{C(s) G(s)}{1 + C(s) G(s)} = \frac{G_o(s)}{1 + G_o(s)}$$

Goal

- Bode plot of $G_o(s) = C(s) G(s)$ can be easily obtained
- Modification of $C(s)$ appears as addition of numerator/denominator factors in the bode plot of $G_o(s)$
- Effect of modification of $C(s)$ on $T(s)$ is not as clear
 \Rightarrow Relate properties of Bode plot of $G_o(s)$ to properties of $T(s)$

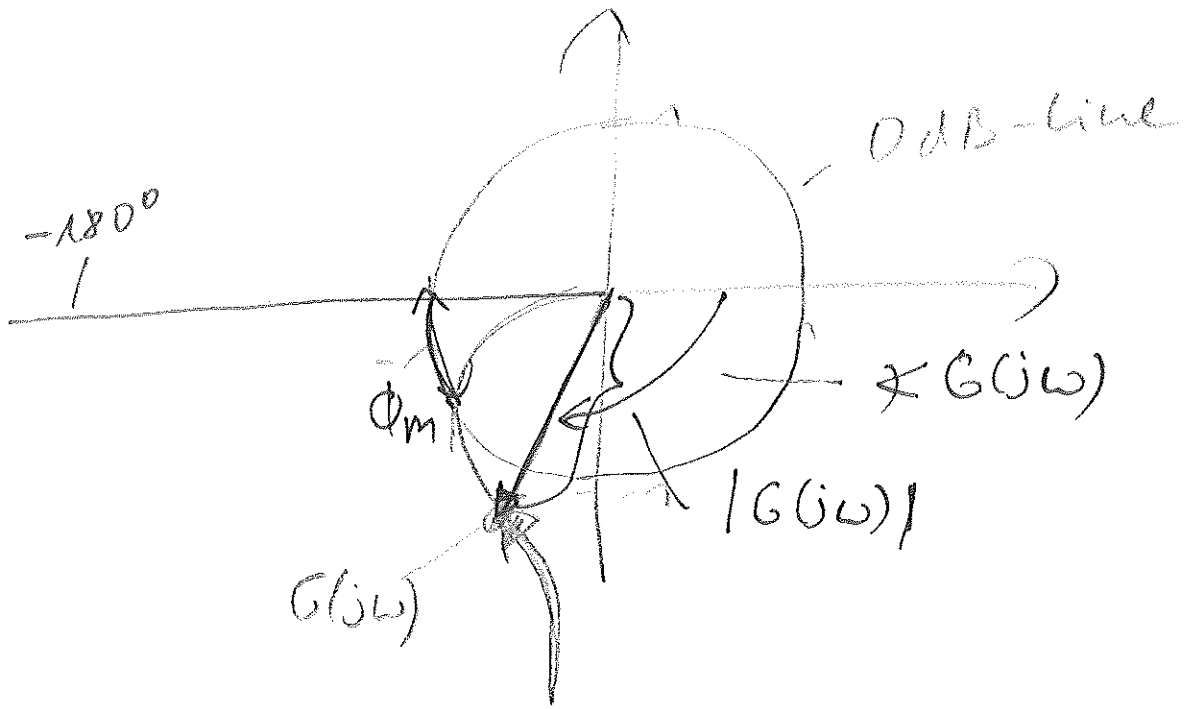
Closed-loop Requirements: Dominant Complex Pole Pair

Approximated Closed-loop Transfer Function

$$T(s) = \frac{K}{1 + 2DTs + T^2s^2}$$

Performance Parameters

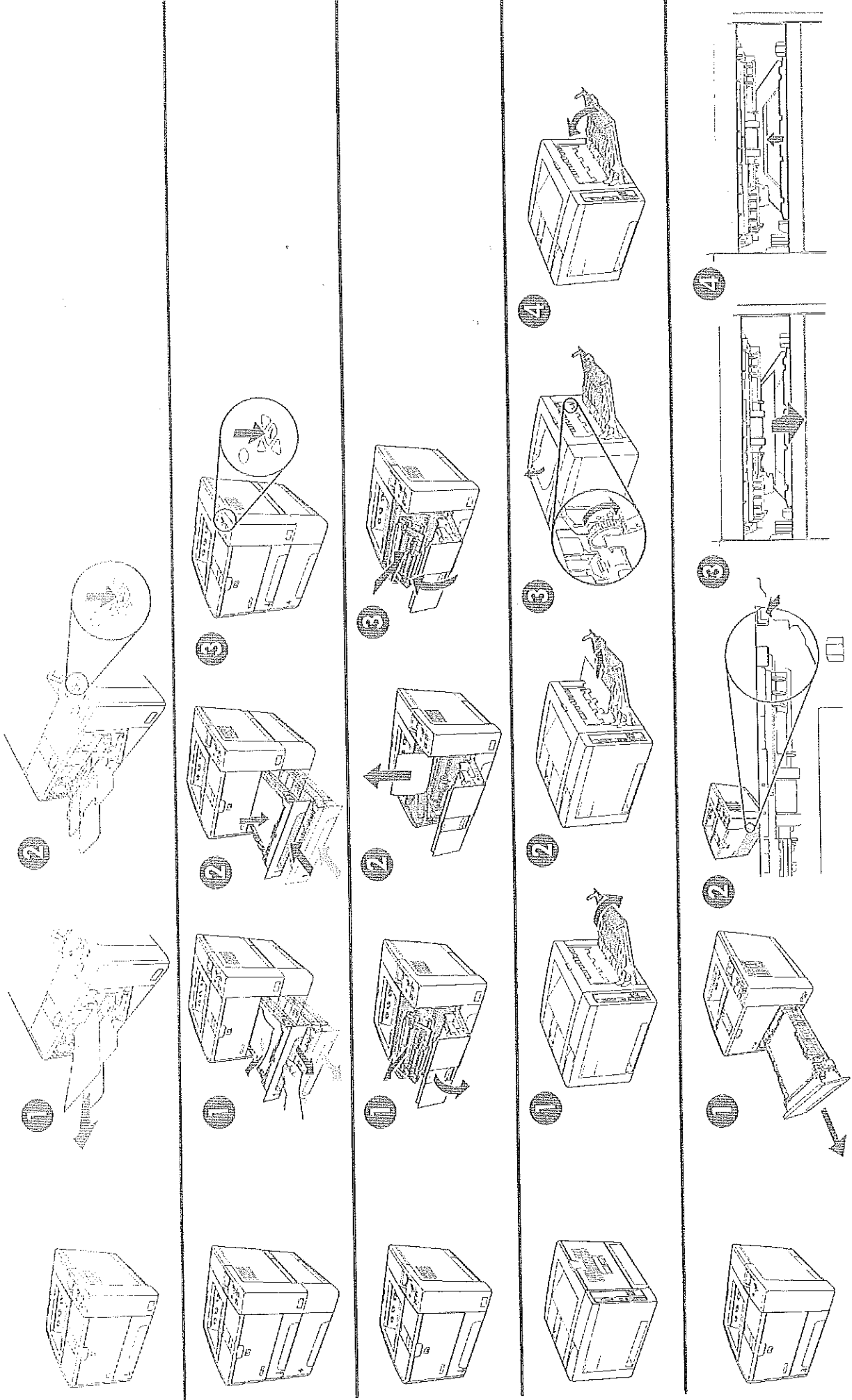
- DC Gain: $K_{DC} = K$
 \Rightarrow Determines final value of the unit step response
- Overshoot: $M_p = e^{-\frac{D\pi}{\sqrt{1-D^2}}}$
- Phase margin: $\Phi_m = \arctan\left(\frac{2D}{\sqrt{-2D^2 + \sqrt{1+4D^4}}}\right)$
 \Rightarrow Determines amplitude of oscillations and relative stability
- Cut-off frequency: $\omega_c = \frac{1}{T} \sqrt{1 - 2D^2 + \sqrt{(1 - 2D^2)^2 + 1}}$
 \Rightarrow Determines speed of response



HP LaserJet P3005 printers



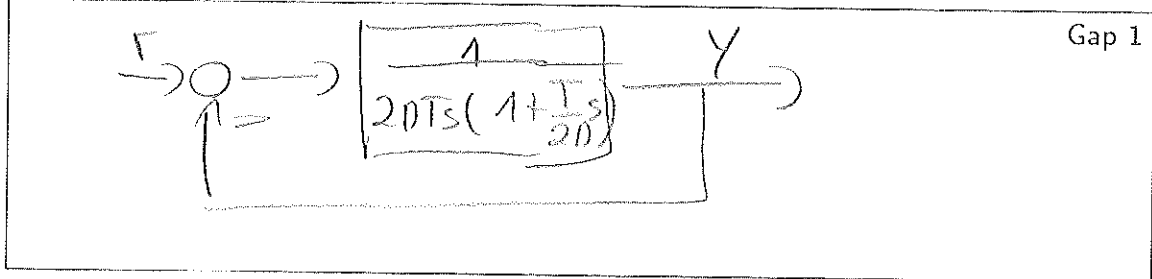
Clear jams 清除卡紙	Eliminer les bourrages Odstranění uvíznutých médií 용지 걸림 해결 Rensa trassel	Staus beseitigen Afhjælp papirstop Fjerne fastkjørt papir 清除卡紙	Eliminazione degli inceppamenti Storingen verhelpen Usuwanie zacięć Kağıt sıkışmalarını giderme	Eliminación de atascos Tukosten poistaminen Limpar atolamentos
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Closed-loop Requirements: Performance Specification

Desired Performance

- DC Gain: $K \approx 1$



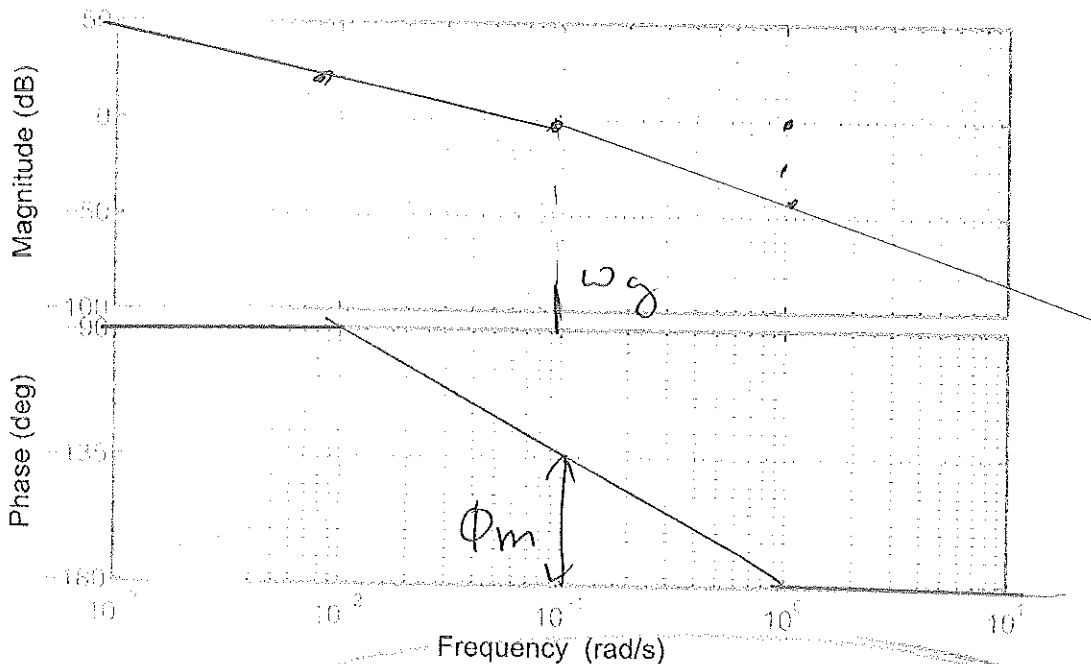
- Open-loop transfer function: $G_o(s) = \frac{1}{2DTs} \frac{1}{(1 + \frac{T}{2D}s)}$
- Sufficient relative stability and small overshoot:
 - \Rightarrow Appropriate choice of D . Generally, $D > 0.4$
- Sufficient speed of response:
 - \Rightarrow Appropriate choice of ω_c

Frequency Response: Open-loop Bode Plot

$D = 0.5, T = 20$

$G_o(s) = \frac{0.1}{s(1 + 10s)}$

Bode Diagram

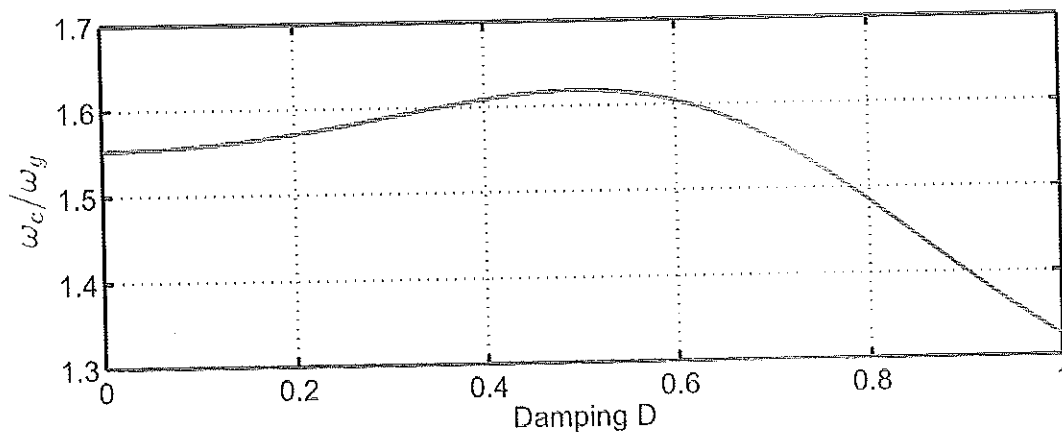


add relation to Nyquist

Closed-loop Requirements: Open-loop Transfer Function

Parameters in Bode Diagram

- Gain crossover frequency: $\omega_g = \frac{1}{T} \sqrt{-2D^2 + \sqrt{1 + 4D^4}}$
 \Rightarrow It holds the $\omega_g \approx \omega_c$



- Phase margin: Φ_m
 \Rightarrow Compute from stability requirement. Usually, $\Phi_m > 40^\circ$

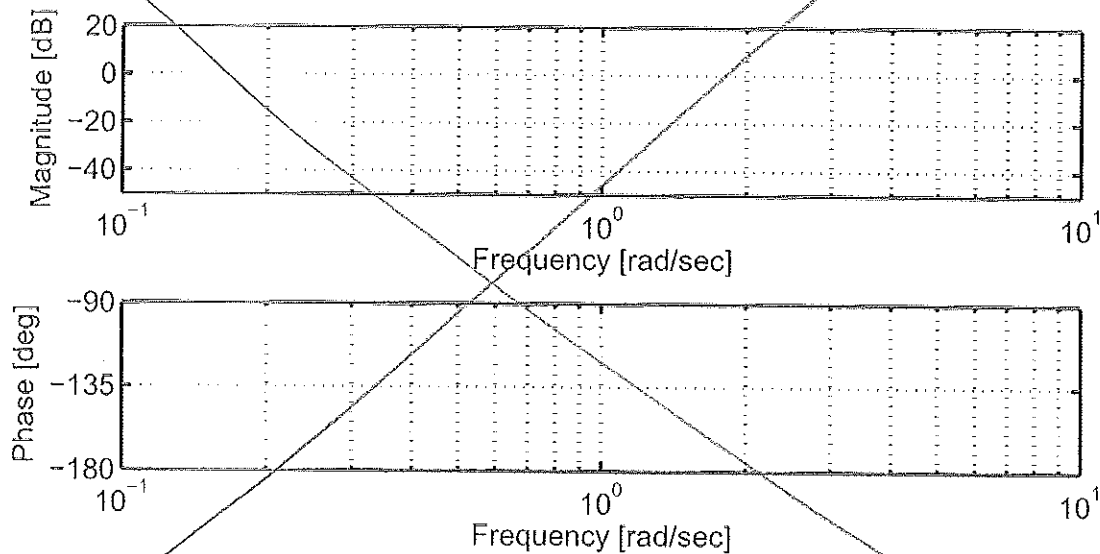
Closed-loop Requirements: Basic Control Design

Procedure

- Determine the closed-loop requirements
 $\Rightarrow \omega_c$ and Φ_m
- Determine the desired open-loop transfer function $G_o(s)$
 $\Rightarrow \omega_g \approx \omega_c$ and Φ_m
- Determine the Bode plot of the plant transfer function $G(s)$
 \Rightarrow Fixed part of the control system
- Design the controller transfer function $C(s)$.
 \Rightarrow Choose the controller poles, zeros and gain such that Bode plot of $C(s)G(s)$ approximately matches $G_o(s)$ around the frequency ω_g

Closed-loop Requirements: Example

not needed

Computation $G(s) = \frac{0.1}{s(s+1)}$ and proportional control $K_p = 10$ 

⇒ Closed-loop poles at $-0.5 \pm 0.87j$

Lead Compensator: Transfer Function

Time-constant Representation

$$C(s) = K_\alpha \frac{1 + Ts}{1 + \alpha Ts}$$

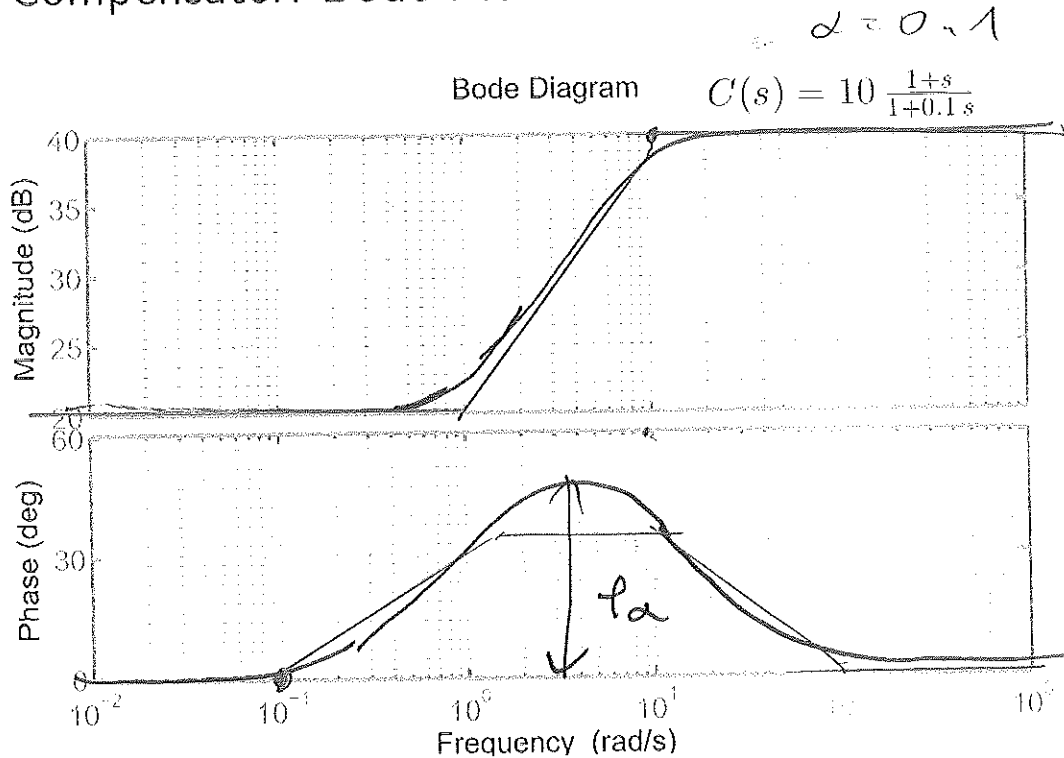
Explanation

- Attenuation factor $0 < \alpha < 1$
- Gain K_α
- Pole at $s = -\frac{1}{\alpha T}$
- Zero at $s = -\frac{1}{T}$

Remarks

- Phase increase (lead) up to a maximum value of $\sin(\varphi_\alpha) = \frac{1-\alpha}{1+\alpha}$
- Frequency of maximum phase lead at $\omega_\alpha = \frac{1}{\sqrt{\alpha}T}$
- Magnitude at ω_α is $|C(j\omega_\alpha)| = \frac{1}{\sqrt{\alpha}} > 1$

Lead Compensator: Bode Plot



Lead Compensator: Usage

Goal

- Reshape frequency response curve to give additional phase lead in order to increase the phase margin

Starting Point

- Plant transfer function $G(s)$
- Lead compensator transfer function $C(s) = K_\alpha \frac{1+Ts}{1+\alpha Ts}$
- Desired phase margin Φ_m
- Steady-state error e_∞

Task

- Determine the parameters K_α , α and T

Lead Compensator: Procedure

- 1. Determine the gain K_α to achieve the static error specification
- 2. Draw a Bode plot of $K_\alpha G(j\omega)$ and determine the phase margin Φ'_m
- 3. Determine the required lead angle $\varphi_\alpha = \Phi_m - \Phi'_m + 10^\circ$

$$\Rightarrow \alpha = \frac{1 - \sin(\Phi_{max})}{1 + \sin(\Phi_{max})}$$
- 4. Choose the gain crossover frequency ω_g such that

$$|K_\alpha G(j\omega_g)|_{dB} = -20 \log\left(\frac{1}{\sqrt{\alpha}}\right)$$

$$\Rightarrow \text{Choose } \omega_\alpha = \omega_g$$
- 5. Evaluate $\omega_\alpha = \frac{1}{\sqrt{\alpha} T}$

$$\Rightarrow T = \frac{1}{\omega_\alpha \sqrt{\alpha}}$$
- 6. Verify if the design fulfills the specified requirements. Go back to step 3. if the requirements are not fulfilled

Lead Compensator: Example

Computation $G(s) = \frac{4}{s(s+2)}$

We want static velocity error is $\frac{1}{20}$ Gap 2
 Phase margin $\Phi_u = 50^\circ$

1. $\lim_{s \rightarrow 0} \frac{1}{1+G(s) \cdot C(s)} \stackrel{\text{Final Value Thm}}{=} \lim_{s \rightarrow 0} \frac{1}{1 + \frac{4}{s(s+2)} \cdot K_\alpha \cdot \frac{1}{s}} = \lim_{s \rightarrow 0} \frac{2s}{2s + 4K_\alpha} \cdot \frac{1}{s} = \frac{1}{2K_\alpha} = \frac{1}{20} \Rightarrow K_\alpha = 10$

2. \rightarrow see bode plot

3. $\Phi'_u \approx 10^\circ$ (from bode plot)
 \Rightarrow we need $\varphi_\alpha = \Phi_u - \Phi'_u + 10^\circ = 40^\circ$
 $\Rightarrow \alpha \approx 0.2$

Lead Compensator: Example

Computation

4. $-20 \log(\frac{1}{\sqrt{a}}) = -7 \text{ dB}$

Gap 3

$\Rightarrow \omega_a = 9 \text{ rad/sec}$

5. $T = \frac{1}{\omega_a \sqrt{a}} = \frac{1}{9 \sqrt{0.2}} \text{ sec} = 0.23 \text{ sec}$

6. Modify Bode plot

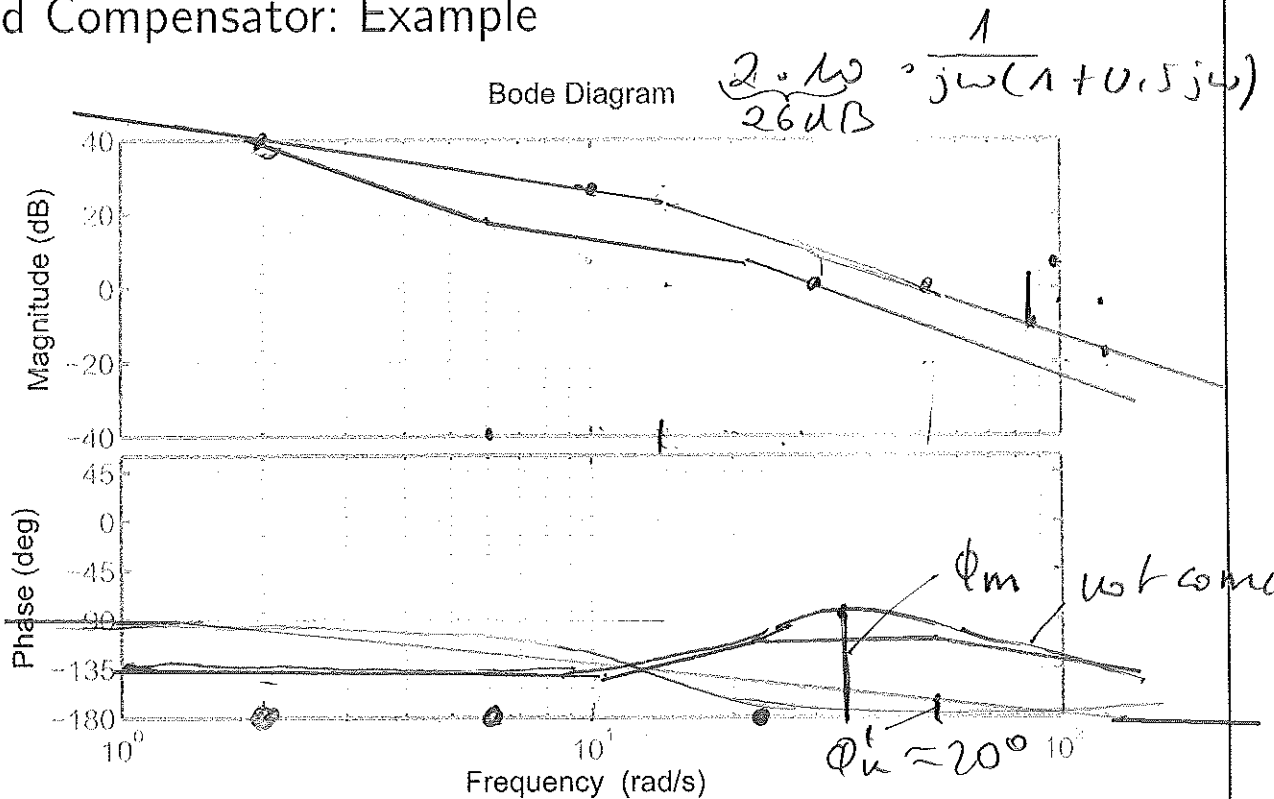
$$C(s) = 10 \cdot \frac{1 + 0.23 s}{1 + 0.045 s}$$

$$\Rightarrow G_o(j\omega) = 200 \cdot \frac{1 + 0.23 j\omega}{1 + 0.045 j\omega} \cdot \frac{1}{j\omega(1 + 0.5 j\omega)}$$

Lead Compensator: Example

shift G to left!

Bode Diagram



\Rightarrow plot plot!

Lag Compensator: Transfer Function

Time-constant Representation

$$C(s) = K_{\beta} \frac{1 + Ts}{1 + \beta Ts}$$

Explanation

- Attenuation factor $\beta > 1$
- Gain K_{β}
- Pole at $s = -\frac{1}{\beta T}$
- Zero at $s = -\frac{1}{T}$

Remarks

- Phase decrease (lag) up to a maximum value of $\sin(\varphi_{\beta}) = \frac{\beta-1}{1+\beta}$
- Frequency of maximum phase lag at $\omega_{\beta} = \frac{1}{\sqrt{\beta} T}$
- Magnitude at ω_{β} is $|C(j\omega_{\beta})| = \frac{1}{\sqrt{\beta}}$

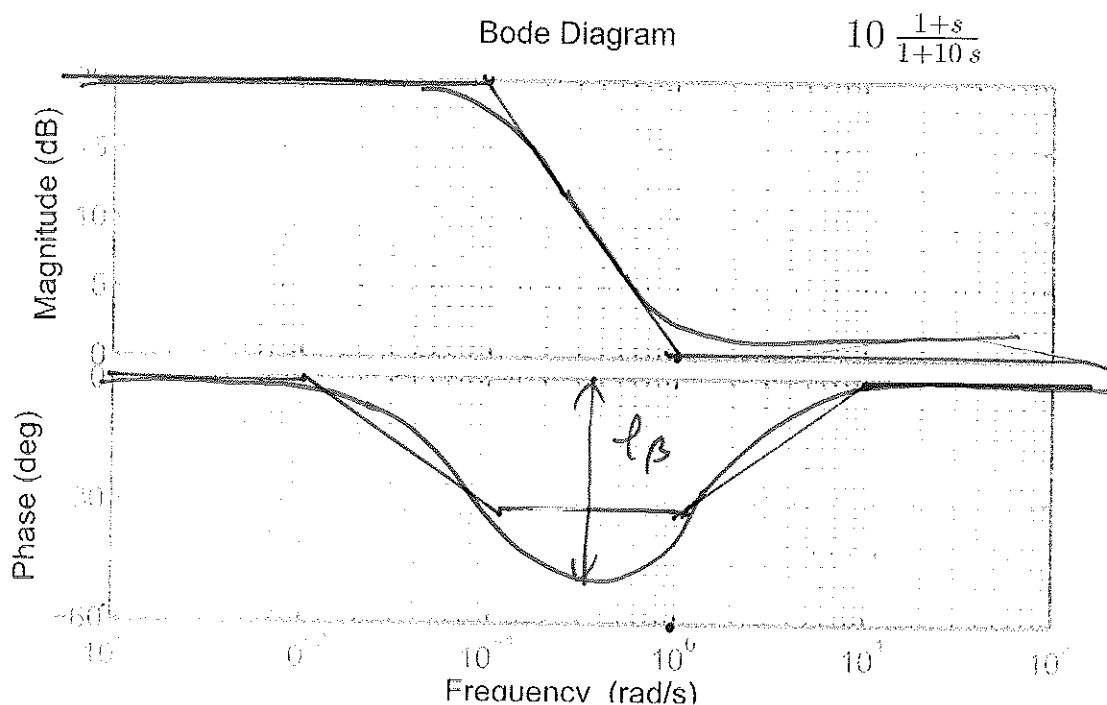
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Lag Compensator: Bode Plot

$$\beta = 10$$



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Lag Compensator: Usage

Goal

- Reshape frequency response curve to give attenuation in the high-frequency range in order to provide enough phase margin

Starting Point

- Plant transfer function $G(s)$
- Lag compensator transfer function $C(s) = K_\beta \frac{1 + Ts}{1 + \beta Ts}$
- Desired phase margin Φ_m
- Steady-state error e_∞

Task

- Determine the parameters K_β , β and T

Lag Compensator: Procedure

- Determine the gain K_β to achieve the static error specification
- Draw a Bode plot of $K_\beta G(j\omega)$ and determine the phase margin Φ'_m
- Find the gain crossover frequency ω_g , where the phase angle is equal to $-180^\circ + \Phi_m + 10^\circ$
- Choose $\frac{1}{T}$ one octave (factor 2) or better one decade (factor 10) below ω_g
 \Rightarrow Phase lag does not affect gain crossover frequency
- Determine β such that $20 \log(\beta) = |K_\beta G(j\omega)|_{dB}$
 \Rightarrow Magnitude curve crosses the 0-dB line at ω_g
- Verify if the design fulfills the specified requirements. Go back to step 3. if the requirements are not fulfilled

Lag Compensator: Example

→ mon to
Home work!

$$\text{Computation } G(s) = \frac{5/4}{(1+10s)(1+4s)}$$

Specifications

Gap 4

- static position error $e_{ss} \approx 1/M$
- phase margin $\phi_m = 45^\circ$

Steps:

$$1. \lim_{s \rightarrow 0} s \cdot \frac{1}{1+G_0} = \lim_{s \rightarrow 0} \frac{1}{1+K_B \cdot \frac{5/4}{(1+10s)(1+4s)}} = \frac{1}{1+5/4 \cdot K_B} = \frac{1}{M} \Rightarrow K_B = 8$$

$$2. K_B G(s) = \frac{10}{(1+10s)(1+4s)}$$

$$3. \text{phase angle } -180^\circ + \phi_m + 180^\circ = -125^\circ \\ \Rightarrow \omega_g \approx 0.3 \text{ rad/sec}$$

Lag Compensator: Example

Computation

Gap 5

$$4. \text{Choose } \frac{1}{T} = 0.03 = 0.1 \cdot \omega_g \Rightarrow T = 33 \text{ sec}$$

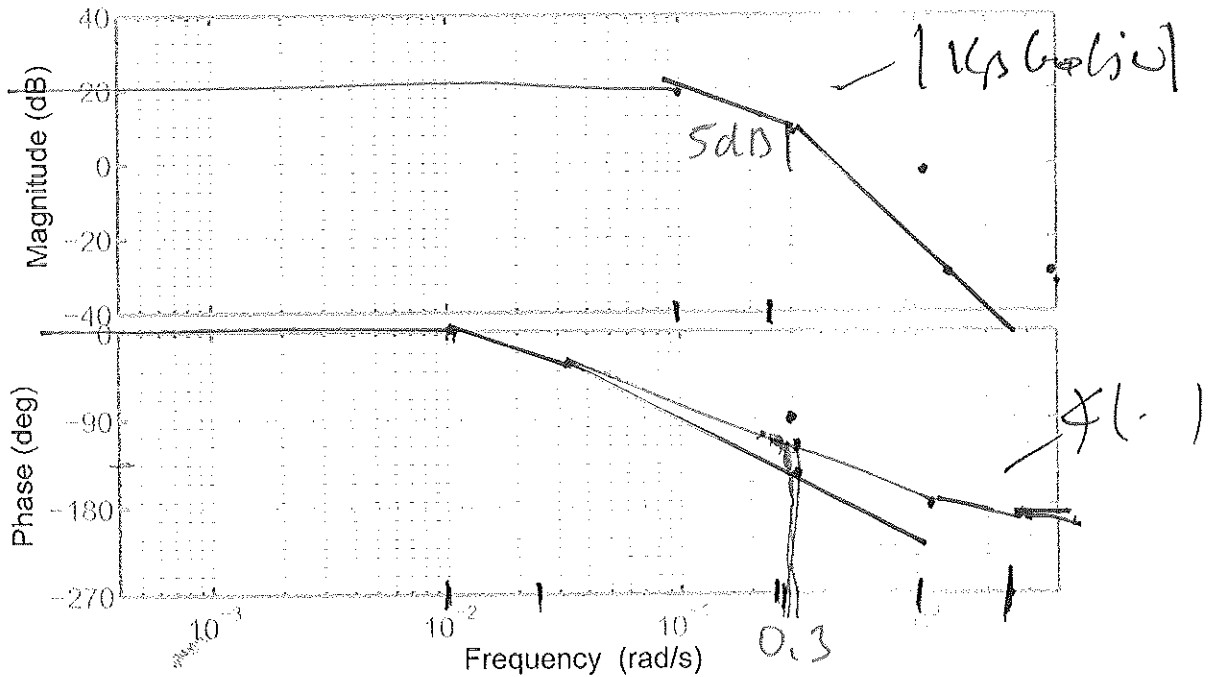
$$5. 20 \log(M) = |K_B G(s \omega_g)| \approx 5 \text{ dB}$$

$$\Rightarrow B = 10^{5/20} \approx 1.8$$

$$\text{Together } C(s) = 8 \cdot \frac{1+33 \cdot s}{1+1.8 \cdot 33 s}$$

Lag Compensator: Example

Bode Diagram



Lag-Lead Compensator: Explanation

Transfer Function

$$C(s) = K \frac{1 + Ts}{1 + \alpha Ts} \frac{1 + Ts}{1 + \beta Ts}$$

- $0 < \alpha < 1$ and $\beta > 1$

Description

- Combines advantages of lead compensator and lag compensator
- No detailed discussion in this lecture
- See for example textbook by Ogata, Section 9-4